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August 2009

FAN5307 High-Efficiency Step-Down DC-DC Converter

Features

- 95% Efficiency, Synchronous Operation
- Adjustable Output Voltage Option: 0.7V to 0.8VIN
- 2.5V to 5.5V Input Voltage Range
- **Customized Fixed Output Voltage Options**
- Up to 300mA Output Current
- Fixed Frequency 1MHz PWM Operation
- High-Efficiency, Power-Save Mode
- 100% Duty Cycle Low Dropout Operation
- Soft Start
- Dynamic Output Voltage Positioning
- 15µA Quiescent Current
- **Excellent Load Transient Response**
- 5-Lead SOT-23 Package
- 6-Lead MLP 3x3mm Package

Applications

- Pocket PCs, PDAs
- Cell Phones
- Battery-Powered Portable Devices
- **Digital Cameras**
- Low Power DSP Supplies

Description

The FAN5307, a high-efficiency, low-noise synchronous PWM current mode and Pulse Skip (Power-Save) mode DC-DC converter, is designed for battery-powered applications. It provides up to 300mA of output current over a wide input range from 2.5V to 5.5V. The output voltage can be either internally fixed or externally adjustable over a wide range of 0.7V to 0.8V_{IN} by an external voltage divider. Custom output voltages are also available. Contact a Fairchild sales representative for customized output voltage options.

At moderate and light loads, pulse skipping modulation is used. Dynamic voltage positioning is applied, and the output voltage is shifted 0.8% above nominal value, for increased headroom during load transients. At higher loads, the system automatically switches to current mode PWM control, operating at 1 MHz. A current mode control loop with fast transient response ensures excellent line and load regulation. In Power-Save mode, the quiescent current is reduced to 15µA to achieve high efficiency and ensure long battery life. In shutdown mode, the supply current drops below 1µA.

The device is available in 5-lead SOT-23 and 6-lead MLP 3x3mm packages.

Ordering Information

Part Number	Operating Temperature Range	V _{out} (V)	Package	© Eco Status	Packing Method
FAN5307S18X	-40 to +85°C	1.8	5-Lead SOT-23	RoHS	Tape and Reel
FAN5307MP18 X	-40 to +85°C	1.8	6-lead 3x3mm Molded Leadless Package (MLP)	Green	Tape and Reel
FAN5307S15X	-40 to +85°C	1.5	5-Lead SOT-23	RoHS	Tape and Reel
FAN5307MP15 X	-40 to +85°C	1.5	6-lead 3x3mm Molded Leadless Package (MLP)	Green	Tape and Reel
FAN5307SX	-40 to +85°C	Adjustable	5-Lead SOT-23	RoHS	Tape and Reel
FAN5307MPX	-40 to +85°C	Adjustable	6-lead 3x3mm Molded Leadless Package (MLP)	Green	Tape and Reel

For Fairchild's definition of Eco Status, please visit: http://www.fairchildsemi.com/company/green/rohs_green.html.

Typical Applications

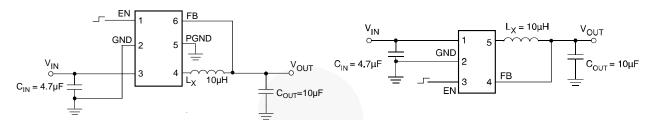


Figure 1. 6-Lead 3x3mm (MLP)

Figure 2. 5-Lead SOT-23

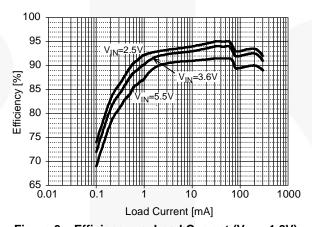


Figure 3. Efficiency vs. Load Current (V_{OUT} =1.8V)

Pin Configuration

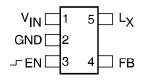


Figure 4. 5-Lead SOT-23

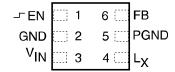


Figure 5. 6-Lead 3x3mm MLP

Pin Definitions

5-Lead SOT-23

<u> </u>	00: 20	
Pin#	Name	Description
1	V_{IN}	Supply Voltage Input.
2	GND	Ground.
3	EN	Enable Input . Logic HIGH enables the chip; logic LOW disables the chip and reduces supply current to <1 μ A. Do not float this pin. If the EN pin is tied to V _{IN} , V _{IN} must be ramped up faster than 5V/ms for V _{OUT} to enter regulation.
4	FB	Feedback Input . In case of fixed-voltage options, connect this pin directly to the output. For an adjustable voltage option, connect this pin to the resistor divider.
5	L _X	Inductor Pin. This pin is connected to the internal MOSFET switches.

6-Lead 3x3mm MLP

Pin#	Name	Description			
1	EN	Enable Input . Logic HIGH enables the chip; logic LOW disables the chip and reduces supply current to <1 μ A. Do not float this pin. If the EN pin is tied to V _{IN} , V _{IN} must be ramped up faster than 5V/ms for V _{OUT} to enter regulation.			
2	GND	Reference Ground.			
3	V_{IN}	Supply Voltage Input.			
4	L _X	nductor Pin. This pin is connected to the internal MOSFET switches			
5	PGND	Power Ground. The internal N-channel MOSFET is connected to this pin.			
6	FB	Feedback Input . In case of fixed-voltage options, connect this pin directly to the output. For an adjustable voltage option, connect this pin to the resistor divider.			

Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

Symbol	Para	Min.	Max.	Unit	
V _{IN}	Supply Voltage		-0.3	6.5	V
VIN	Input Voltage on PVIN and An	y Other Pin	GND-0.2	V _{IN} +0.3	V
0	Thermal Resistance ⁽¹⁾	Junction-to-Case, SOT-23		130	°C/W
θις	Thermal Resistance	Junction-to-Tab, MLP 3x3		8	
T∟	Lead Soldering Temperature,		+260	°C	
T _{STG}	Storage Temperature		-65	+150	°C
ESD ⁽²⁾	Human Body Model, JESD22-	A114	4		kV
ESD	Charged Device Model, JESD	22-C101	1		٨V

Notes

- Junction-to-ambient thermal resistance, θ_{JA}, is a strong function of PCB material, board thickness, thickness and number of copper planes, number of via used, diameter of via used, available copper surface, and attached heat sink characteristics.
- 2. Using Mil Std. 883E, method 3015.7 (Human Body Model) and EIA/JESD22C101-A (Charged Device Model).

Recommended Operating Conditions

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to Absolute Maximum Ratings.

Symbol	Parameter	Min.	Тур.	Max.	Unit
V _{IN}	Supply Voltage Range	2.5		5.5	V
V _{OUT}	Output Voltage Range, Adjustable Version	0.7		0.8V _{IN}	V
I _{OUT}	Output Current			300	mA
L	Inductor ⁽³⁾		10		μΗ
C _{IN}	Input Capacitor ⁽³⁾		4.7		μF
C _{OUT}	Output Capacitor ⁽³⁾		10		μF
T _A	Operating Ambient Temperature Range	-40		+85	°C
TJ	Operating Junction Temperature Range	-40		+125	°C

Note:

3. Refer to the Applications section for details.

Electrical Characteristics

 V_{IN} =2.5V to 5.5V, I_{OUT} =200mA, EN= V_{IN} , C_{IN} =4.7 μ F, C_{OUT} =22 μ F, L_X =10 μ H, T_A =-40°C to +85°C, unless otherwise noted. Typical values are at T_A =25°C.

Symbol	Pa	rameter	Conditions	Min.	Тур.	Max.	Units
V _{IN}	Input Voltage			2.5		5.5	V
ΙQ	Quiescent Curre	nt	I _{OUT} =0mA, Device is not switching		15	30	μΑ
I _{SD}	Shutdown Suppl	y Current	EN=GND		0.1	1.0	μΑ
V _{ENH}	Enable High Inpo	ut Voltage		1.3			V
V _{ENL}	Enable Low Inpu	it Voltage				0.5	V
I _{EN}	En Input Bias Cu	ırrent	EN=V _{IN} or GND		0.01	0.10	μΑ
	DMOC On Desig	tanaa	V _{IN} =V _{GS} =3.6V	1 200	530	690	0
Б	PMOS On Resis	tance	V _{IN} =V _{GS} =2.5V		670	850	mΩ
R _{DS-ON}	NIMOO O D :	. \	V _{IN} =V _{GS} =3.6V		430	540	
	NMOS On Resis	tance	V _{IN} =V _{GS} =2.5V		530 660		mΩ
I _{LIM}	P-channel Curre	nt Limit	2.5V < V _{IN} < 5.5V	400	520	700	mA
I _{Ikg_(N})	N-channel Leaka	age Current	V _{DS} =5.5V		0.1	1.0	μA
I _{Ikg_(P)}	P-channel Leaka	age Current	V _{DS} =5.5V		0.1	1.0	μA
	Switching Frequency			800	1000	1200	kHz
R _{LINE}	Line Regulation		V _{IN} =2.5 to 5.5, I _{OUT} =10mA		0.16		% / V
D	Load	6-Lead MLP	$\begin{array}{l} 100\text{mA} \leq I_{\text{OUT}} \\ \leq 300\text{mA} \end{array}$		0.0014		% / mA
R _{LOAD}	Regulation	5-Lead SOT-23	$\begin{array}{l} 100\text{mA} \leq I_{OUT} \\ \leq 300\text{mA} \end{array}$		0.0022		% / mA
		6-Lead MLP	$\begin{aligned} &V_{\text{IN}}\text{=}2.5 \text{ to } 5.5\text{V},\\ &0\text{mA} \leq I_{\text{OUT}} \leq 300\text{mA} \end{aligned}$	-4		4	
V _{OUT}	Output Voltage Accuracy	5-Lead SOT-23	$\begin{aligned} &V_{\text{IN}}\text{=}2.5 \text{ to } 4.5\text{V},\\ &0\text{mA} \leq I_{\text{OUT}} \leq 300\text{mA} \end{aligned}$	-4		4	%
		5-Lead 501-23	$\begin{split} &V_{\text{IN}}\text{=}2.5 \text{ to } 5.5\text{V},\\ &0\text{mA} \leq I_{\text{OUT}} \leq 300\text{mA} \end{split}$	-5		4	
I _{LEAK}	Leakage Curren	t into Pin SW	$\begin{aligned} &V_{IN}{>}V_{OUT},\\ &0V \leq V_{SW} \leq V_{IN} \end{aligned}$		0.1	1.0	μA
I _{LEAK_R}	Reverse Leakag	e Current into Pin SW	V _{IN} =Open, EN=GND, V _{SW} =5.5		0.1	1.0	μA

Electrical Characteristics for Adjustable Version

 $V_{IN}\!\!=\!\!2.5V \text{ to } 5.5V, \, I_{OUT}=200 \text{mA}, \, EN=V_{IN}, \, C_{IN}\!\!=\!\!4.7 \mu F, \, C_{OUT}\!\!=\!\!22 \mu F, \, L_X\!\!=\!\!10 \mu H, \, T_A\!\!=\!\!25^{\circ}C.$

Symbol	Parameter	Min.	Тур.	Max.	Units
V_{FB}	Feedback Voltage		0.5		V

Electrical Characteristics for Fixed V_{OUT}=1.8 Version

 V_{IN} =2.5V to 5.5V, I_{OUT} =200mA, EN= V_{IN} , C_{IN} =4.7 μ F, C_{OUT} =22 μ F, L_X =10 μ H, T_A =-40°C to +85°C, unless otherwise noted. Typical values are at T_A =25°C.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
Vpfm_pwm	PFM to PWM Transition	$\begin{split} &V_{\text{IN}}\!\!=\!3.7V,T_{A}\!\!=\!25^{\circ}\text{C},\\ &0.1\text{mA} \leq I_{\text{OUT}} \leq 300\text{mA} \end{split}$			72	mV
	Voltage ⁽⁴⁾	$\begin{aligned} &V_{\text{IN}}\text{=-}4.2\text{V}, \text{ T}_{\text{A}}\text{=-}25^{\circ}\text{C}, \\ &0.1\text{mA} \leq I_{\text{OUT}} \leq 300\text{mA} \end{aligned}$			12	
V _{OUT_TRANS}	Output Voltage During Mode Transition ^(5,6)		1.70		1.93	V
V _{OUT_CLAMP}	Over-Voltage Clamp Threshold	Includes Line, Load, Load Transients, and Temperature		1.878	1.930	V

Note:

- 4. Transition voltage is defined as the difference between the output voltage measured at 0.1mA (PFM mode) and 300mA (PWM mode), respectively.
- 5. See Figure 6.
- 6. These limits also apply to any mode transition caused by any kind of load transition within specified output current range.

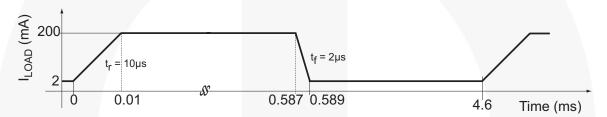


Figure 6. Load Transient Response Test Waveform

Typical Performance Characteristics

 $T_A=25$ °C, $C_{IN}=C_{OUT}=10\mu F$, $L=10\mu H$, $V_{OUT}=1.8V$, unless otherwise noted.

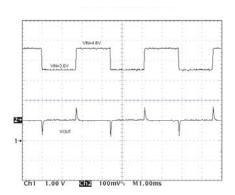


Figure 7. Line Transient Response

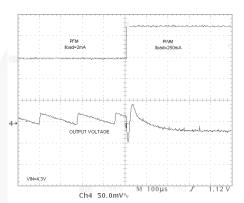


Figure 9. Load Transient Response

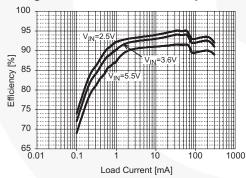


Figure 11. Efficiency vs. Load Current (V_{OUT}=1.8V)

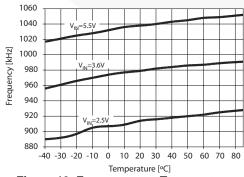


Figure 13. Frequency vs. Temperature

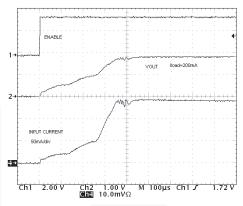


Figure 8. Startup

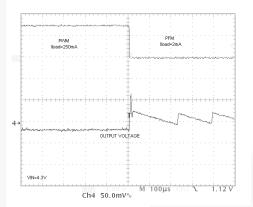


Figure 10. Load Transient Response

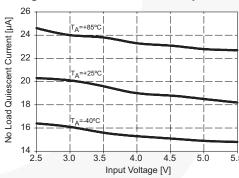


Figure 12. No-Load Quiescent Current vs. VIN

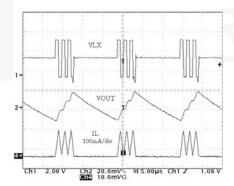


Figure 14. Power Save (PRM) Mode Operation

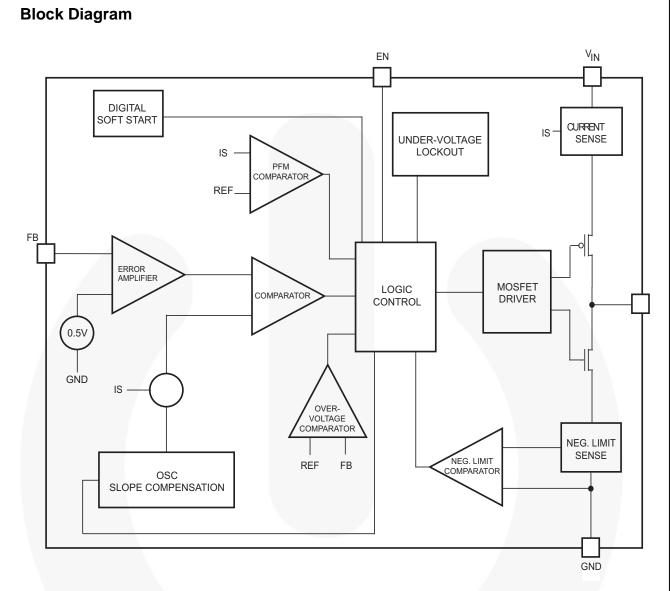


Figure 15. Block Diagram

Detailed Operation Description

The FAN5307 is a step-down converter operating in a current-mode PFM/PWM architecture with a typical switching frequency of 1MHz. At moderate to heavy loads, the converter operates in pulse-width-modulation (PWM) mode. At light loads, the converter enters a power-save mode (PFM pulse skipping) to keep the efficiency high.

PWM Mode

In PWM mode, the device operates at a fixed frequency of 1MHz. At the beginning of each clock cycle, the P-channel transistor is turned on. The inductor current ramps up and is monitored via an internal circuit. The P-channel switch is turned off when the sensed current causes the PWM comparator to trip when the output voltage is in regulation or when the inductor current reaches the current limit (set internally to typically 520mA). After a minimum dead time, the N-channel transistor is turned on and the inductor current ramps down. As the clock cycle is completed, the N-channel switch is turned off and the next clock cycle starts.

FM (Power-Save) Mode

As the load current decreases and the peak inductor current no longer reaches the typical threshold of 80mA, the converter enters pulse-frequency-modulation (PFM) mode. In PFM mode, the device operates with a variable frequency and constant peak current, reducing the quiescent current to a minimum and maintaining high efficiency at light loads. As soon as the output voltage falls below a threshold, set at 0.8% above the nominal value, the P-channel transistor is turned on and the inductor current ramps up. The P-channel switch turns off and the N-channel turns on as the peak inductor current is reached (typical 140mA).

The N-channel transistor is turned off before the inductor current becomes negative. At this time, the P-channel is switched on again, starting the next pulse. The converter continues these pulses until the high threshold is reached (typically 1.6% above nominal value). A higher output voltage in PFM mode gives additional headroom for the voltage drop during a load transient from light to full load. The voltage overshoot during this load transient is minimized due to active regulation during turning on the N-channel rectifier switch. The device stays in sleep mode until the output voltage falls below the low threshold. FAN5307 enters PWM mode as soon as the output voltage can no longer be regulated in PFM with constant peak current.

100% Duty Cycle Operation

As the input voltage approaches the output voltage and the duty cycle exceeds the typical 90%, the converter turns the P-channel transistor continuously on. In this mode, the output voltage is equal to the input voltage minus, the voltage drop across the P-channel transistor:

$$V_{OUT} = V_{IN} - I_{LOAD} \times (R_{DSON} + R_L), \text{ where}$$
 (1)

 R_{DSON} = P-channel switch on resistance I_{LOAD} = Output current

R_L = Inductor DC resistance

Soft Start

The FAN5307 has an internal soft-start circuit that limits the inrush current during start-up. This prevents possible voltage drops of the input voltage and eliminates the output voltage overshoot. The soft-start is implemented as a digital circuit, increasing the switch current in four steps to the P-channel current limit (520mA). Typical start-up time for a 10µF output capacitor and a load current of 200mA is 500µs.

Short-Circuit Protection

Switch peak current is limited, cycle by cycle, to a typical value of 520mA. In an output voltage short circuit, the device operates at minimum duty cycle; therefore, the average input current is typically 100mA.

Application Information

Adjustable Output Voltage Version

The output voltage for the adjustable version is set by the external resistor divider, as shown below:

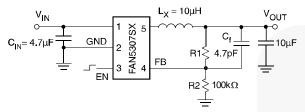


Figure 16. External Resistor Divider

calculated as:

$$V_{OUT} = 0.5V \times \left[1 + \frac{R_1}{R_2}\right]$$
 (2)

To reduce noise sensitivity, R1 + R2 should not exceed $1M\Omega$.

Inductor Selection

The inductor parameters directly related to device performance are saturation current and DC resistance. The FAN5307 operates with a typical inductor value of 10µH. The lower the DC resistance, the higher the efficiency. For saturation current, the inductor should be rated higher than the maximum load current, plus half of the inductor ripple current, calculated as:

$$\Delta I_{L} = V_{OUT} \times \frac{1 - (V_{OUT} / V_{IN})}{L \times f}$$
 (3)

where:

f = Switching frequency

L = Inductor value

 ΔI_L = Inductor ripple current

Table 1. Recommended Inductors

Inductor Value	Vendor Part Number		Perfor- mance
		CDRH5D28-100	
10	Sumida	CDRH5D18-100	Highest
10μH		CDRH4D28-100	Efficiency
	Murata	LQH66SN100M01L	
6.8µH		CDRH3D16-6R8	
	0	CDRH4D18-100	
10µH	Sumida	CR32-100	Smallest
		CR43-100	Solution
	Murata	LQH4C100K04	
	Cooper Bussmann	CTX01-17327	

Input Capacitor Selection

For best performances, a low-ESR input capacitor is required. A ceramic capacitor of at least $4.7\mu F$, placed as close to the input pin of the device, is recommended.

Output Capacitor Selection

The FAN5307's switching frequency of 1MHz allows the use of a low-ESR ceramic capacitor with a value of $10\mu F$ to $22\mu F$. This provides low output voltage ripple. In power-save mode, the output voltage ripple is independent of the output capacitor value and the ripple is determined by the internal comparator thresholds. The typical output voltage ripple at light load is 1% of the nominal output voltage.

Table 2. Recommended Capacitors

Capacitor Value	Vendor	Part Number
4.7µF		JMK212BY475MG
	Taiyo Yuden	JMK212BJ106MG
40 .		JMK316BJ106KL
10μF	TDK	C12012X5ROJ106K
	IDK	C3216X5ROJ106M
22µF	Murata	GRM32DR60J226K

PCB Layout Recommendations

The inherently high peak currents and switching frequency of the power supplies require careful PCB layout design. Use wide traces for the high-current path and place the input capacitor, the inductor, and the output capacitor as close as possible to the integrated circuit terminals. For the adjustable version, the resistor divider should be routed away from the inductor to avoid electromagnetic interference.

The 6-lead MLP version of the FAN5307 separates the high-current ground from the reference ground; therefore, it is more tolerant to the PCB layout design and shows better performance.

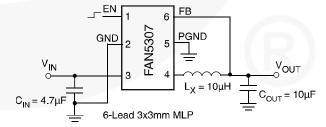
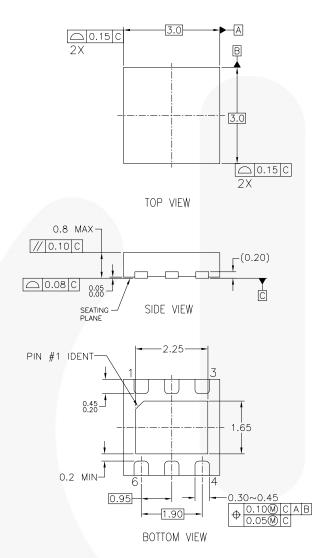
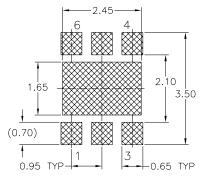


Figure 17. Possible Layout

Physical Dimensions





RECOMMENDED LAND PATTERN

NOTES:

- A. CONFORMS TO JEDEC REGISTRATION MO-229, VARIATION WEEA, DATED 11/2001
- B. DIMENSIONS ARE IN MILLIMETERS.
- C. DIMENSIONS AND TOLERANCES PER ASME Y14.5M, 1994

MLP06DrevA

Figure 18. 3x3mm 6-Lead Molded Leadless Package (MLP)

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Physical Dimensions

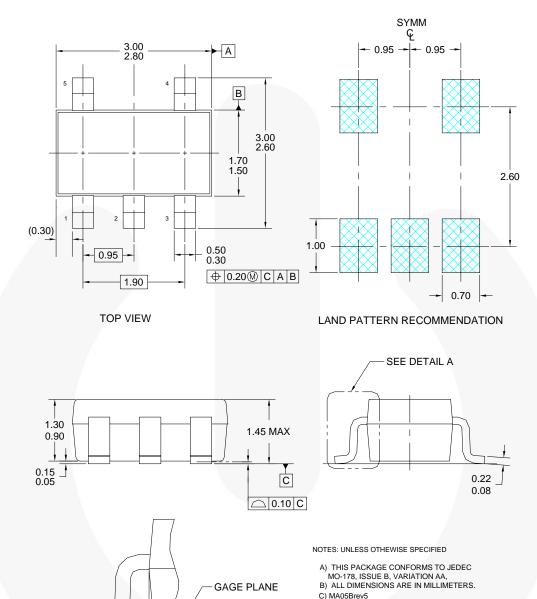


Figure 19. 5-Lead SOT-23 Package

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SEATING PLANE

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CTL™ Gmax™
Current Transfer Logic™ GTO™
EcoSPARK® IntelliMAX™
EfficientMax™ ISOPLANAR™
EZSWTCH™ MegaBuck™
MICROCOUPLER™
MicroFET™

F® MicroPak™
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 FastvCore™

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